

1 Numerical Methods

1.1 Fourier Transforms of Bessel Functions (6 units)

This project assumes only material contained in Part IA and Part IB core courses. Other than that, the project is self contained (although the Part II courses on Numerical Analysis, Further Complex Methods and Asymptotic Methods may provide relevant but non-essential background).

1 Introduction

Bessel's equation of order n is the linear second-order equation

$$x^2 y'' + xy' + (x^2 - n^2)y = 0. \quad (1)$$

Bessel functions of the first kind are solutions of (1) which are finite at $x = 0$. They are usually written $J_n(x)$.

Question 1 Investigate (1) for $n = 0, 1, 4$ using a Runge–Kutta (or similar) method commencing the integration for a strictly positive value of x and a number of different values of y and y' of your choice. You may employ a library routine to solve (1); for example if using MATLAB you can employ the built-in `ode45` routine. Integrate forwards and backwards in x for a few such initial conditions, plotting y . Describe what you observe, and illustrate any notable behaviour using appropriate plots.

Now try starting at $x = 0$. What happens, and why?

Question 2 The series solution for $J_n(x)$ is

$$J_n(x) = \sum_{r=0}^{\infty} \frac{(-1)^r (\frac{1}{2}x)^{2r+n}}{r!(n+r)!}. \quad (2)$$

Write a program to sum a truncation of this series. Plot $J_n(x)$ for $n = 0, 1, 4$ for a range of x , e.g., for $0 \leq x \leq 100$. Discuss your choice of truncation, and identify a range of x for which this summation method is not accurate and explain why.

2 The Discrete Fourier Transform

The Fourier Transform $\hat{F}(k)$ of a function $F(x)$ may be defined as

$$\hat{F}(k) = \int_{-\infty}^{+\infty} F(x) \exp(-2\pi i k x) dx. \quad (3)$$

If $F(x)$ is a function which is only appreciably non-zero over a limited range of x , say $0 < x < X$, then it is possible to approximate $\hat{F}(k)$ by means of finite sums. Suppose

$$F_r = F(r\Delta x) \quad \text{for } r = 0, \dots, (N-1), \quad \text{where } \Delta x = X/N. \quad (4)$$

An approximation to (3), known as the Discrete Fourier Transform (DFT), is

$$\hat{\mathcal{F}}_s = \frac{X}{N} \sum_{r=0}^{N-1} F_r \omega_N^{-rs}, \quad \text{where } \omega_N = e^{2\pi i/N}. \quad (5)$$

The *exact* inverse of (5) is

$$F_r = \frac{1}{X} \sum_{s=0}^{N-1} \hat{\mathcal{F}}_s \omega_N^{rs}. \quad (6)$$

In order to deduce the relationship between the $\hat{\mathcal{F}}_s$ and $\hat{F}(k)$, we first note from (5) that $\hat{\mathcal{F}}_s$ represents values of the Fourier Transform spaced by the “wavenumber” interval Δk , where

$$\Delta k = 1/X. \quad (7)$$

Also $\hat{\mathcal{F}}_s$ is periodic in s with period N ; this corresponds to a “wavenumber” periodicity

$$K = N\Delta k = N/X = 1/\Delta x. \quad (8)$$

Now it is to be expected that (5) will fail to approximate to (3) when the exponential function oscillates significantly between sample points, that is when

$$|k| \gtrsim \frac{1}{2\Delta x} = \frac{1}{2}K. \quad (9)$$

This, together with its periodicity, suggests that $\hat{\mathcal{F}}_s$ will be related to $\hat{F}(k)$ by

$$\hat{\mathcal{F}}_s \cong \begin{cases} \hat{F}(s\Delta k) & s = 0, \dots, \frac{1}{2}N - 1, \\ \hat{F}(s\Delta k - K) & s = \frac{1}{2}N, \dots, N - 1. \end{cases} \quad (10)$$

Thus (6) is an approximation to

$$F(x) \cong \int_{-K/2}^{+K/2} \hat{F}(k) \exp(2\pi i k x) dk. \quad (11)$$

Because of the periodicity, the $\hat{\mathcal{F}}_s$ are usually thought of as a series with $s = 0, \dots, N - 1$, the upper half being mentally re-positioned to correspond to negative “wavenumber”. Note that if $F(x)$ is real, and $*$ denotes a complex conjugate, then

$$\hat{F}(k) = \hat{F}^*(-k). \quad (12)$$

Question 3 Carefully discuss under what limiting conditions for both N and X (possibly after a suitable change in origin in x), does the DFT tend to the Fourier Transform?

3 The Fast Fourier Transform

The Fast Fourier Transform (FFT) method provides an efficient way to evaluate the DFT. This method involves efficient evaluation of sums of the form

$$\lambda_s = \sum_{r=0}^{N-1} \mu_r \omega_N^{\sigma r s}, \quad s = 0, \dots, N - 1, \quad \sigma = \pm 1, \quad (13)$$

where N is an integer and the μ_r are a known sequence. The “fast” in FFT requires N to be a power of a small prime, or combination of small primes; for simplicity we will assume that $N = 2^M$.

A brief outline of the FFT method is given in the Appendix. However, it is not necessary to understand the implementation details, since you may use the one-dimensional Fast Fourier Transform function `fft` in MATLAB, SCIPY, NUMPY, or an equivalent routine in any other package. Alternatively you may write your own routine (however do **not** simply compute the series (5) and (6); you will not receive credit if you do not use the FFT method.).

Note that the `fft` algorithm will work for any value of N , but the performance (in terms of compute time) is best when N is a power of 2.

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Question 4 Show analytically that if $F(x)$ is a real even function and

$$I_1 = \int_0^X F(x) \exp(-2\pi i k x) dx, \quad I_2 = \int_{-X}^{+X} F(x) \exp(-2\pi i k x) dx, \quad (14a)$$

then

$$\operatorname{Im}(I_2) = 0, \quad \operatorname{Re}(I_2) = 2\operatorname{Re}(I_1). \quad (14b)$$

With the definitions of §2 and §3, the FFT algorithm is ideally suited to approximating I_1 rather than I_2 . Hence if an approximation to I_2 is desired, an approximation to I_1 could first be calculated, and then the relations (14b) could be used. If this procedure for calculating I_2 is adopted, and $F_N \neq F_0$, explain why F_0 should be replaced by $\frac{1}{2}(F_0 + F_N)$ before calculating the DFT. What is the equivalent result to (14b) if $F(x)$ is a real odd function?

Question 5 Using a FFT code, and the results of question 4, find numerically the Fourier Transform of $J_n(x)$:

$$\hat{J}_n(k) = \int_{-\infty}^{+\infty} J_n(x) \exp(-2\pi i k x) dx. \quad (15)$$

Compare it with the theoretical formula

$$\hat{J}_n(k) = 2(-i)^n (1 - 4\pi^2 k^2)^{-1/2} T_n(2\pi k), \quad (16)$$

where $T_n(\mu)$ is the Chebyshev polynomial of order n defined by

$$T_n(\mu) = \begin{cases} \cos n\theta, & \mu = \cos \theta; \\ 0, & |\mu| > 1. \end{cases} \quad (17)$$

To obtain $J_n(x)$, you may either devise a method of your own (e.g., a combination of questions 1 and 2), or you may use the MATLAB procedure `besselj` or a similar function from another package.

You should obtain results for $n = 0, 1, 2, 4$, and 8 . Choose sufficient points in the transform to adequately resolve the functions.

Plots of $J_n(x)$ for a few representative values of n should be included in your write-up. You should also include plots of \hat{J}_n and $\hat{\mathcal{J}}_n$ on the same graph. Choose a range of k which allows you to see the detailed behaviour in the interval $-1 \leq \pi k \leq 1$.

Comment on your results and discuss their accuracy. Discuss how the FFT deals with any values of k which might be expected from the theoretical result to give problems. You should also describe the effects of varying N and X ; in particular you should **systematically** examine how the numerical errors change as N and/or X are varied, e.g. in the light of your answer to question 3.

You should also find a way to demonstrate from your computational results how the execution time necessary to calculate the transform varies with N , and how this compares with the theoretical prediction. *Hint:* given the speed of current computers, timing a *single* run of your program is likely to be dominated by start/end overheads.

Appendix: The Fast Fourier Transform

The Fast Fourier Transform (FFT) technique is a quick method of evaluating sums of the form

$$\lambda_r = \sum_{s=0}^{N-1} \mu_s \omega_N^{\sigma r s}, \quad r = 0, \dots, N-1, \quad \sigma = \pm 1, \quad (18)$$

where N is an integer, μ_s is a known sequence and $\omega_N = e^{2\pi i/N}$. The “fast” in FFT depends on N being a power of a small prime, or combination of small primes; for simplicity we will assume that $N = 2^M$. Write

$$\lambda_r \longleftrightarrow \mu_s, \quad r, s = 0, \dots, N-1 \quad (19)$$

to denote that (18) is satisfied. Introduce the half-length transforms

$$\left. \begin{aligned} \lambda_r^E &\longleftrightarrow \mu_{2s} \\ \lambda_r^O &\longleftrightarrow \mu_{2s+1} \end{aligned} \right\} \quad r, s = 0, \dots, \frac{1}{2}N-1; \quad (20)$$

then it may be shown that

$$\left. \begin{aligned} \lambda_r &= \lambda_r^E + \omega_N^{\sigma r} \lambda_r^O \\ \lambda_{r+N/2} &= \lambda_r^E - \omega_N^{\sigma r} \lambda_r^O \end{aligned} \right\} \quad r = 0, \dots, \frac{1}{2}N-1. \quad (21)$$

Hence if the half-length transforms are known, it costs $\frac{1}{2}N$ products to evaluate the λ_r .

To execute an FFT, start from N vectors of unit length (i.e., the original μ_s). At the s th stage, $s = 1, 2, \dots, M$, assemble 2^{M-s} vectors of length 2^s from vectors of length 2^{s-1} – this “costs” $2^{M-s} \times \frac{1}{2}(2^s) = 2^{M-1} = \frac{1}{2}N$ products for each stage. The complete discrete Fourier transform has been formed after M stages, i.e., after $O(\frac{1}{2}N \log_2 N)$ products. For $N = 1024 = 2^{10}$, say, the cost is $\approx 5 \times 10^3$ products, compared to $\approx 10^6$ products in naive matrix multiplication.

A description and short history of the FFT are given in Chapter 12 of the book *Numerical Recipes* by Press *et al.*